

## Research Article

# Design of a PID Controller using a Fictitious Exogenous Signal for a Fluctuation System

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### ABSTRACT

The fictitious reference iterative tuning method is used to design Proportional–Integral–Derivative (PID) controllers to achieve desired performance for industrial machines. However, the desired control performance cannot be obtained by this method, if the system characteristics changed in steady state. In this paper, design method of a PID controller using the Fictitious Exogenous Signal is proposed. The change of system characteristics is considered as an impulse-like virtual disturbance. The Fictitious Exogenous Signal to calculate PID parameters is obtained from a set of operational data. The effectiveness of the proposed method has been verified by numerical examples.

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## 1. INTRODUCTION

PID controllers has been used in a lot of industrial machines. Appropriate PID parameters are necessary for a desired performance. However, the parameters are difficult to be decided by trial and error. Therefore, the Chien–Hrones–Reswick (CHR) method, a conventional method, is widely used. In this method, the proportional gain, the integral gain and the differential gain are determined based on the parameters of the controlled system, respectively. However, this method can be applied only when the system parameters are known, and desired response characteristics cannot necessarily be obtained in some cases.

Recently, the Fictitious Reference Iterative Tuning (FRIT) method which calculates PID parameters directly using the closed-loop data obtained by a set of operational data attracts attention [1]. The method can be applied, even when the characteristics of the system are unknown. The fictitious reference input is calculated from system input and output in the operational data, and input to the reference model. This method designs the optimum controller by adjusting PID parameters to let the output of the reference model be close to the output in the operational data. However, for many industrial machines, especially machines that operated by human, the system characteristics change when the operating environment changes. In such cases, it is necessary to redesign the controller after the system fluctuation.

In this paper, the controlled system is assumed to be switched in steady state, and the controller is constituted as a PID controller. The PID parameters before the system switching are set by the FRIT method. After the system switching, a method that setting the PID parameters by the Fictitious Exogenous Signal is proposed.

Specifically, the system fluctuation in steady state is regarded as an impulse-like virtual disturbance is input. At this time, the transfer function from the disturbance to the control error is constructed. The optimum controller is designed by adjusting the PID parameters so that the characteristics of the closed-loop system approaches the reference model. Genetic Algorithm (GA) is used to calculate the PID parameters. Moreover, the effectiveness of the proposed method has been verified by numerical examples.

## 2. CONTROL SYSTEM

### 2.1. Control Objective

A switching system in which the system changing occurs in steady state is shown in Figure 1 as the control objective. System1 is initial state, and system2 is the state after switching. Switching of the system is assumed to occur at a known time. Controller1 and controller2 are controllers for these systems, respectively. PID parameters are set in the controllers.

### 2.2. Control Law

The controllers for System1 and System2 use the following I-PD controller [Equation (1)].

$$\Delta u(k) = K_i e(k) - K_p \Delta y(k) - K_d \Delta^2 y(k) \quad (1)$$

$u(k)$ ,  $e(k)$  and  $y(k)$  are control input, control error and system output.  $K_p$ ,  $K_i$  and  $K_d$  are proportional gain, integral gain and differential gain, respectively.  $\Delta$  is a differencing operator, defined by  $\Delta := 1 - z^{-1}$ .

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Furthermore, the control error  $e(k)$  is calculated as the following Equation (2).  $r(k)$  is the reference value.

$$e(k) := r(k) - y(k) \quad (2)$$

### 2.2.1. FRIT method

The controller1 for system1 is set as  $K_{p1}$ ,  $K_{i1}$ , and  $K_{d1}$  using the FRIT method. Here,  $K_{p1}$ ,  $K_{i1}$ , and  $K_{d1}$  are proportional gain, integral gain and differential gain, respectively. The controller parameters are calculated directly by using a set of operational data, and the block diagram is shown in Figure 2.  $u_0(k)$  and  $y_0(k)$  in the closed-loop data are system input and output of a set of initial operational data.  $\tilde{y}(k)$  is the output of a reference model  $G_m(z^{-1})$ .  $\tilde{r}(k)$  indicates the fictitious reference input of the reference model  $G_m(z^{-1})$ , and calculated from a set of operational data ( $u_0(k)$ ,  $y_0(k)$ ) and Equation (1).  $G_m(z^{-1})$  and  $\tilde{r}(k)$  are defined as the following Equations (3)–(5).

$$\tilde{y}(k) = G_m(z^{-1})\tilde{r}(k) = \frac{z^{-1}P(1)}{P(z^{-1})}\tilde{r}(k) \quad (3)$$

$$P(z^{-1}) = 1 + p_1z^{-1} + p_2z^{-2} \quad (4)$$

$$\tilde{r}(k) = y_0(k) + \frac{\Delta u_0(k) + K_{p1}\Delta y_0(k) + K_{d1}\Delta^2 y_0(k)}{K_{i1}} \quad (5)$$

Here,  $p_1$  and  $p_2$  are designed by the following Equation (6) [2].

$$\left. \begin{aligned} p_1 &= -2\exp\left(-\frac{\rho}{2\mu}\right)\cos\left(\frac{\sqrt{4\mu-1}}{2\mu}\rho\right) \\ p_2 &= \exp\left(-\frac{\rho}{\mu}\right) \\ \rho &= \frac{T_s}{\sigma} \\ \mu &= 0.25(1-\delta) + 0.51\delta \end{aligned} \right\} \quad (6)$$

$\sigma$  relates to the rising characteristic of the control, and  $\delta$  is a parameter relating to the attenuation characteristics (it can be set by the

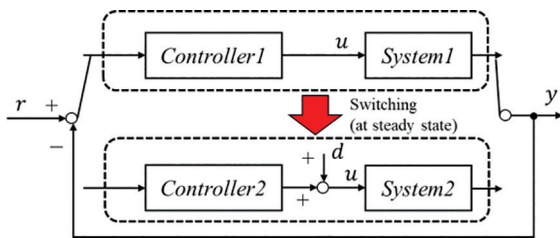


Figure 1 | Switching system.

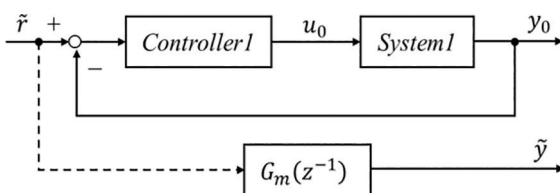


Figure 2 | Block diagram of the FRIT method.

designer, arbitrarily). It is desirable that  $\delta$  is set between  $0 \leq \delta \leq 1$ . When  $\delta = 0$ , the Binominal model response is shown, and when  $\delta = 1$ , the Butterworth model response is shown.

### 2.2.2. Fictitious Exogenous Signal method

The controller2 for system2 is set as  $K_{p2}$ ,  $K_{i2}$  and  $K_{d2}$  by a method using the Fictitious Exogenous signal. Here,  $K_{p2}$ ,  $K_{i2}$  and  $K_{d2}$  are proportional gain, integral gain and differential gain, respectively. The controller parameters are calculated directly by using a set of operational data. The method using the Fictitious Exogenous Signal is similar to the FRIT method, and block diagram is shown in Figure 3. The reference model  $G_{md}(z^{-1})$  means the desired closed-loop transfer characteristics from disturbance to control error.  $\tilde{d}(k)$  is the input, and  $\tilde{e}(k)$  is the output. The  $\tilde{d}(k)$  is a Fictitious Exogenous Signal, and the output of the reference model  $G_{md}(z^{-1})$  is  $\tilde{e}(k)$  when the input is  $\tilde{d}(k)$ . The Fictitious Exogenous Signal  $\tilde{d}(k)$  is expressed by the following Equation (7) [3] from a set of initial operational data ( $u_0(k)$ ,  $y_0(k)$ ) measured in advance and Equation (1). Here, the system fluctuation is regarded as an impulse-like virtual disturbance  $\tilde{d}(k)$  was input. At this time, since  $\tilde{d}(k)$  is a virtual value, the initial operational data that it is necessary to calculate the  $\tilde{d}(k)$  is the same as a set of operational data ( $u_0(k)$ ,  $y_0(k)$ ) measured in advance.  $u_c(k)$  is calculated by the controller.  $\tilde{d}(k)$  is calculated by the following Equation (7).

$$\begin{aligned} \tilde{d}(k) &= u_0(k) - u_c(k) \\ &= u_0(k) - (u_c(k-1) + K_{i2}e_0(k) - K_{p2}\Delta y_0(k) - K_{d2}\Delta^2 y_0(k)) \end{aligned} \quad (7)$$

The reference model  $G_{md}(z^{-1})$  can be expressed by the following Equation (8) using the  $G_m(z^{-1})$  in the FRIT method.

$$G_{md}(z^{-1}) = G_m(z^{-1})\left(\frac{K_{i2}}{\Delta}\right)^{-1} \quad (8)$$

Therefore, output  $\tilde{e}(k)$  of the reference model  $G_{md}(z^{-1})$  is calculated by the following Equation (9).

$$\tilde{e}(k) = G_{md}(z^{-1})\tilde{d}(k) = -\frac{z^{-1}P_d(1)}{P_d(z^{-1})}\left(\frac{K_{i2}}{\Delta}\right)^{-1}\tilde{d}(k) \quad (9)$$

The setting method of  $P_d(z^{-1})$  is the same as the FRIT method, and parameters  $\sigma$  and  $\delta$  can be set different values from system1.

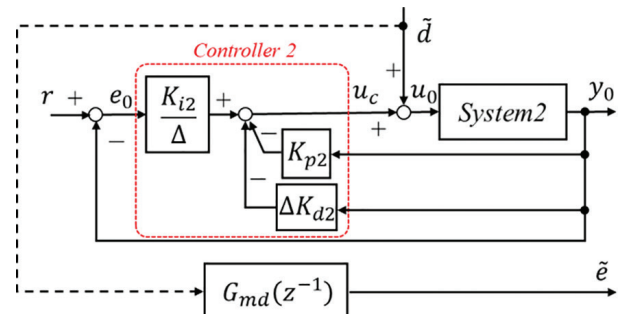


Figure 3 | Block diagram of the proposed method using the Fictitious Exogenous Signal.

### 2.3. Genetic Algorithm

The GA was used to minimize the value of the evaluation function [4,5]. The evaluation function is defined by the following Equations (10) and (11).

$$\begin{aligned} & \text{[[System1]]} \\ & J_1 = \frac{1}{2} \sum_1^{N_1} (y_0(k) - \tilde{y}(k))^2 \end{aligned} \quad (10)$$

$$\begin{aligned} & \text{[[System2]]} \\ & J_2 = \frac{1}{2} \sum_1^{N_2} (e_0(k) - \tilde{e}(k))^2 \end{aligned} \quad (11)$$

$N_1$  and  $N_2$  indicate the total number of steps of operational data in the system1 and the system2, respectively. In the system1, a desired response characteristic can be obtained by defining the difference between the reference model output  $\tilde{y}(k)$  and  $y_0(k)$  as an evaluation function and setting PID parameters to minimize the evaluation value. When the evaluation value is low, the fitness is high and PID parameters are better. The system2 is also the same.

First, initial individuals  $K_p$ ,  $K_i$  and  $K_d$  as genes are randomly generated, and initial evaluation is performed. Individual with high fitness are considered elite, and carry over to the next generation. Next, crossover, tournament selection and mutation are performed based on the evaluation values. These procedures are repeated in each generation, and PID parameters are obtained as the gene of the individual with the highest fitness in the final generation. In this paper, the GA was calculated using Matlab under the condition of 200 individuals and 200 generations.

## 3. NUMERICAL SIMULATION

To verify the effectiveness of the proposed method, a switching system of the first order lag systems was set as the following Equations (12) and (13) and a simulation was performed.

$$\begin{aligned} & \text{[[System1]]} \\ & G_1 = \frac{1}{1+50s} e^{-2s} \end{aligned} \quad (12)$$

$$\begin{aligned} & \text{[[System2]]} \\ & G_2 = \frac{1}{1+60s} e^{-2s} \end{aligned} \quad (13)$$

### 3.1. Conventional Method (CHR Method)

The following Equations (14) and (15) were calculated by discretizing the equations above at the sampling time  $T_s = 1$  s, and are obtained, respectively. Here, the Gaussian white noise with mean 0 and variance 0.01 was given by  $\xi(k)$ . Furthermore, in 250 steps, the system was switched from the system1 to the system2. The effectiveness of the proposed method was verified. Reference is  $r(k) = 10$ .

$$\begin{aligned} & \text{[[System1]]} \\ & y(k) = 0.980y(k-1) + 0.0990u(k-3) + \xi(k) \end{aligned} \quad (14)$$

$$\begin{aligned} & \text{[[System2]]} \\ & y(k) = 0.984y(k-1) + 0.0165u(k-3) + \xi(k) \end{aligned} \quad (15)$$

First, a simulation result using PID parameters calculated by the CHR method was shown in Figure 4. The PID parameters are shown in Table 1.

In Figure 4, it took long time to follow the Reference  $r(k)$ , and the responsiveness of the system1 is not good. The fixed PID parameters corresponding to the system1 has particularly small proportional gain  $K_p$ . Therefore, input  $u(k)$  is moderate at the switching time, and the output variation after switching to the system2 is large and recovery time is long. The control performance is not good.

### 3.2. Proposed Method

Figure 5 shows the control result of the proposed method. The PID parameters for the system1 are calculated using the FRIT method, and the PID parameters for the system2 are calculated by adjustment method using the Fictitious Exogenous Signal.  $u_0(k)$  and  $y_0(k)$  that obtained by the CHR method were used as a set of the operational data in order to calculate the PID parameters for the system1 and the system2. Table 2 shows the PID parameters. As shown in Figure 5, the response of the system1 is faster than the conventional method, and the desired output was obtained. In addition, the output variation after switching to the system2 is less than 1/4 as compared with the conventional method, and good control performance is obtained. The PID parameters for system2 is larger than that for system1. Therefore, the response of input  $u(k)$  is faster and output variation can be suppressed.

It was confirmed that the responsiveness after the system switching was improved by considering the system fluctuation

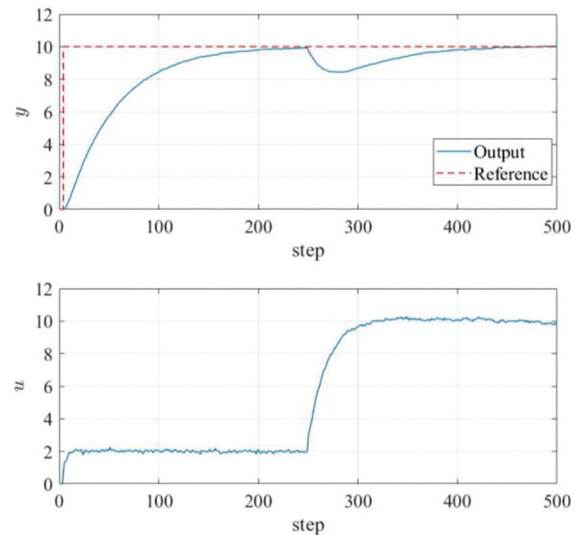
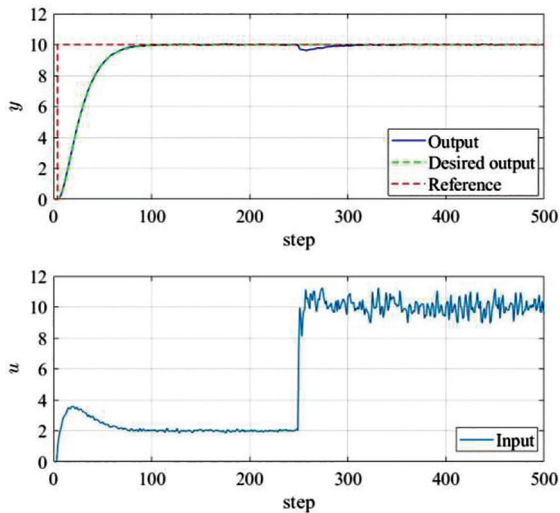


Figure 4 | Simulation result by the CHR method.

Table 1 | PID parameters calculated by the CHR method

	$K_p$	$K_i$	$K_d$
Systems1 and 2	3.0	0.06	3.0



**Figure 5** | Simulation result using the FRIT method and the Fictitious Exogenous Signal.

**Table 2** | PID parameters calculated by the FRIT method and Fictitious Exogenous Signal

	$K_p$	$K_i$	$K_d$
System1	1.60	0.07	3.19
System2	18.4	0.73	20

as a disturbance input and applying the adjustment law using the Fictitious Exogenous Signal.

## 4. CONCLUSION

In this paper, a method of controller adjustment law using the Fictitious Exogenous Signal for a system that switched characteristics

## Authors Introduction

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in steady state was proposed. In the proposed method, the system fluctuation was regarded as a disturbance input, and the transfer function from the disturbance  $d(k)$  to the control error  $e(k)$  was constructed as the reference model  $G_{md}(z^{-1})$ . The Fictitious Exogenous Signal calculated from a set of operational data was input to the reference model  $G_{md}(z^{-1})$ . It was confirmed that a controller with a good disturbance responsiveness can be obtained by adjusting the PID parameters to approximate the reference model output  $\tilde{e}(k)$  to the control error  $e_0(k)$ .

In the future, the proposed method will be applied to an actual machine to confirm the effectiveness.

## CONFLICTS OF INTEREST

The authors declare they have no conflicts of interest.

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